

The development of LLRF system for STCF

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CONTENTS 1. Introduction.

2. LLRF for Linac.

3. LLRF for storage ring.

4. Summary.

/01 Introduction

- The overview of STCF.
- The RF system of STCF.

1.1 The overview of STCF





Center-of-mass energy	/: <mark>2-7 Gev</mark> (-charm th	reshold)
Designed brightness: >	> 0.5×10 ³⁵ cm ⁻² s ⁻¹	
_ line	ear injector	ing completely symmetric
Physical design: - sto	orage ring - positive	and nagative electron beam energy is equal
_ dai	mping ring L only one	e collision point
The	2024 International Wo	orkshop on Future Tau Charm Facilities

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A1 Administrator, 2024/1/12

1.1 The overview of STCF





Parameters		
Beam current(A)	2	
$\mathbf{D}_{\mathbf{a}\mathbf{a}\mathbf{m}} = \mathbf{a}_{\mathbf{a}\mathbf{a}\mathbf{v}}(\mathbf{C}_{\mathbf{a}}\mathbf{V})$	2(storage ring)	
Beam energy(Gev)	3.5(Linac)	
DE frequency (MHz)	497.5 (storage ring)	
Kr frequency(MHZ)	2998 (Linac)	
Harmonic number	1024	
Revolution frequency (kHz)	486.08	

Linear injector:

- electron gun.
- beaming system.
- · linear acceleration sections.
- positron source.
- electron bypass drift section.
- positron damping ring. **Storage ring:**
- electron storage ring.
- positron storage ring.

1.2 The RF system of STCF





- **low-level control system:** control and maintain the phase and amplitude stability of RF station during beam travelling;
- reference signal distribution system: provide reference signals;
- **microwave transmission system:** include highpower waveguide attenuator/phase shifter, waveguide transmission segment, input coupler, coaxial absorption load and so on;
- pulse/continuous power source system: include S-band klystron and focusing coil, high-voltage power supply, power source monitoring system and the solid state modulator;
- **microwave acceleration structure:** include accelerator tube, a pre-buncher and a buncher. The travelling-wave accelerating tube consists of an accelerating unit and a coupler.

/02 LLRF for Linac

- The schematic of the LLRF for Linac(@2998 MHz, pulsed).
- Former research basis of 2856 MHz LLRF for NSRL.
- The following work of LLRF for STCF Linac.

2.1 The schematic of the LLRF for Linac(@2998 MHz, pulsed)

RF system (@2998 MHz)

The STCF requires the construction of electron injectors and positive injectors, and its microwave system needs to provide S-band (2998MHz) acceleration structure and microwave power. (Accelerator pre-research project has been launched in Nov. 2023)

To ensure the steady infusion of energy, we develop a 2998MHz low-level RF system, including signal source, frequency synthesizer, signal processor.





The LLRF system(2856M) operating at HLS II

The digital board in LLRF

- 6 channels 125MS/s 16bit ADC(2 channels reserved);
- 2 channels 250MS/s 16bit DAC;
- 6 channels down-converters;
- 1 channel s-band vector modulator;
- Adjustable attenuator (31 dB).



The offline workbench

Hardware

- Signal source: generate REF signal (2856 MHz) for Linac RF system and frequency synthesizer.
- Frequency synthesizer: generate LO (2829.56 MHz) and CLK signal (105.78 MHz).
- **Signal processor**: monitor & control the solid-state amplifier, klystron and cavity.



Graphic user interface (GUI):

Developed based on phoebus; Implement remote control.

Application software:

Amplitude and phase calculation; Inter-pulse feedback/pulse-to-pulse feedback; Adaptive feedforward; System exception handling.

EPICS IOC:

Provide driver talk to the hardware; Connect EPICS record with the physical hardware data; Provide process value to the upper level.







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Main technical specification				
Number of channels	6			
Channel isolation	65 dB			
Input signal range	15 dBm			
The rising edge	50 ns			
The phase noise	19.17 fs			
Short term amplitude stability(RMS)	0.010%			
Short term phase stability(RMS)	0.0076 °			
RF input attenuation	0~31.75 dB in 0.25 dB			
range	steps			

	Jitter(fs) 10Hz~10MHz	Frequency stability (ppm/24hours)	Amplitude stability(dbm)RMSE
105.78MHz	45.82	0.00945	0.0200
449.56MHz	41.18	0.00667	0.0174
476MHz	38.53	0.00630	0.0150
2829.56MHz	19.70	0.00071	0.0048
2856MHz	18.05	0.00175	0.0255



Long-term phase stability of closed loop: 0.10° (RMS).



Since May. 2022, three self-developed LLRF processors have replaced the No. 6,7, and 8th MTCA based processors in HLS II. They have been running smoothly and reliably over one year and a half, without any failure, which ensuring the stable operation of HLS II under the top-off injection mode.



Hardware aspects:

- Design a signal source which provides the 2998MHz reference signal and a frequency synthesizer which provides the ADC sampling signal(CLK) and local signal(LO);
- Design an IF processor which can generate pulses with repetition rate of 50Hz;
- Design the microwave reference line and phase calibration system, which ensures the consistency among different LLRF systems.

Software aspects:

- > Use the Debian OS to replace the centos system(longer update cycle, higher long-term stability);
- Use the package EPICS 7.0 instead of EPICS 3.1(less bugs);
- Use the PYDM instead of Phoebus(open source, higher development efficiency).

/[] LLRF for storage ring

- The schematic of the LLRF(@497.5MHz,CW).
- Modeling and simulation of LLRF.
- The design of closed-loop control.

3.1 The schematic of the LLRF(@497.5MHz,CW)



3.1 The schematic of the LLRF(@497.5MHz,CW)



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reduce the observed impedance of the beam.

3.2 Modeling and simulation of LLRF



17



The closed-loop transfer function of the feedback system, $L_{\rm dir}$ is the open loop transfer function of the direct feedback loop, and $L_{\rm comb}$ is the open loop transfer function of the comb filter feedback loop:

$$H_{\text{meas}}(s) = \frac{L_{dir}(s)}{1 - (1 + L_{surf}(s))L_{tin}(s)}$$

Where σ is the damping time of the high-frequency cavity

$$\sigma = \omega_{rf} / (2Q_l)$$

$$L_{dir}(s) = \frac{2\sigma s}{s^2 + 2\sigma s + \omega_r^2} H_{lead}(s) G_d e^{i\phi_d - (s - i\omega_{rf})T_d}$$

$$L_{comb}(s) = H_{comb}(s) H_{eq}(s) G_c e^{i\phi_c - (s - i\omega_{rf})T_c}$$



Simulation of the direct feedback loop(@ 500 MHz)





The suppression of specific synchronous oscillation frequency





The function(time domain) of the FIR filter is:

$$C_n = \sin(2\pi\nu(n-1) - \phi)$$

Where n is the filter order, v is the tune of the storage ring, and φ is the tunable phase.

Under the current tune of 0.0097, the longitudinal feedback filter of 105 order is shown in figure (left).

3.3 The design of closed-loop control



Frequency control

The relationship between the detuning Angle($\Delta \phi$) of the high-frequency cavity and the detuning frequency(Δf):

$$\tan\Delta\varphi = -2\frac{\Delta f}{f_0} * Q_L$$



The frequency control loop is generally composed of a digital phase detector module, control tuning motor (tuner) and piezo.

- Digital phase detection module: Calculate the phase difference between pickup and the input signal.
- **Control tuned motor (large range, slow speed) :** By changing the axial length of the accelerator cavity, the equivalent capacitance of the accelerator cavity is changed, and the resonant frequency is changed.
- **Piezoelectric ceramics (small range, fast speed) :** Piezoelectric ceramics use high pressure to change the length of the piezoelectric ceramics, which causes the axial length of the acceleration cavity to be stretched or compressed.



4.1 Summary



Current progress:

- Complete the hardware and software of the LLRF for NSRL(@2856 MHz, S-band, pulsed).
- Design a series of feedback loops for the storage ring, including direct control loop, comb filter feedback loop, mode feedback loop, modulator jitter feedforward loop, frequency control loop.
- Simulation of the feedback loops for the storage ring.

Future plans:

- > Design and complete a new LLRF system(@2998 MHz) for the Linac on the basis of former work;
- Redesign the frequency of CLK and LO;
- Change the data transmission mode to accommodate the high repetition frequency mode;
- Add phase drift calibration function to increase system control accuracy.
- Design and complete the hardware and software for the storage ring LLRF(@497.5 MHz, CW);
- More complete modeling and simulation to support our design;
- > Design a new system to achieve reference line synchronization and phase drift calibration.

Thanks!